

Dual Low-Dropout Linear Regulators with RESET or Low-Noise Output in UCSP or QFN

General Description

The MAX8530/MAX8531 offer the benefits of low-dropout voltage and ultra-low power regulation in subminiaturized UCSP and QFN packages with an integrated microprocessor reset circuit (MAX8530 only). The devices operate from a 2.5V to 6.5V input and deliver up to 200mA and 150mA outputs with low dropout of 100mV (typ) at 100mA. Designed with an internal P-channel MOSFET pass transistor, the supply current is kept at a low 130 μ A (with both LDOs on), independent of the load current and dropout voltage. Other features include short-circuit protection and thermal-shutdown protection.

The MAX8530 provides microprocessor open-drain, active-low reset output to monitor OUT1, eliminating external components and adjustments. The MAX8530 asserts a 100ms (min) RESET signal when OUT1 drops below 87% of the nominal output voltage. The MAX8531 includes a reference bypass pin for low output noise $(40\mu V_{RMS})$.

Both devices include a logic-controlled shutdown input and are available in 6-bump UCSP and 6-pin thin QFN exposed pad packages.

Applications

Cellular and Cordless Phones PDAs and Palmtop Computers Notebook Computers

Digital Cameras

PCMCIA Cards

Wireless LAN Cards

Hand-Held Instruments

Features

- ♦ Guaranteed 200mA Output Current for OUT1
- ♦ Guaranteed 150mA Output Current for OUT2
- ♦ Low 100mV (typ) Dropout at 100mA Load for both LDOs
- ♦ Open-Drain, Active-Low 100ms (min) Reset Timer (MAX8530)
- ♦ Low 40µV_{RMS} Output Noise (MAX8531)
- ♦ Low 130µA Operating Supply Current
- ♦ <1µA Shutdown Current
- ♦ Thermal-Overload and Short-Circuit Protection
- **♦ Output Current Limit**
- ♦ Tiny Packages Available 1.16 x 1.57 x 0.6mm UCSP (3 x 2 Grid) 3mm x 3mm Thin QFN

Ordering Information

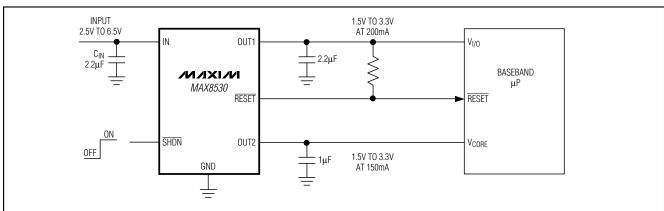
PART	TEMP RANGE	OUT_ VOLTAGE	PIN- PACKAGE	
MAX8530EBTxy*	-40°C to +85°C	1.5V to 3.3V	6 UCSP	
MAX8530ETTxy*	-40°C to +85°C	1.5V to 3.3V	6 Thin QFN-EP**	
MAX8531EBTxy*	-40°C to +85°C	1.5V to 3.3V	6 UCSP	
MAX8531ETTxy*	-40°C to +85°C	1.5V to 3.3V	6 Thin QFN-EP	

^{*}xy = Output voltage code (see the Output Voltage Selector Guide).

Pin Configurations appear at end of data sheet.

Output Voltage Selector Guide appears at end of data sheet

Typical Operating Circuit



MIXIM

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

^{**}EP = Exposed pad.

ABSOLUTE MAXIMUM RATINGS

IN, $\overline{\text{SHDN}}$, $\overline{\text{RESET}}$, BP to GND0.3V to +7V OUT1, OUT2 to GND0.3V to (V _{IN} + 0.3V) Output Short-Circuit DurationIndefinite Continuous Power Dissipation (TA = +70°C) 6-Bump UCSP (derate 3.9mW/°C above +70°C)308mW	Operating Temperature Range40°C to +85°C Junction Temperature+150°C Storage Temperature Range65°C to +150°C 6-Bump+ UCSP Solder Profile(Note 1) 6-Lead QFN Lead Temperature (10s)+300°C
6-Lead OFN (derate 24 4mW/°C above +70°C) 1951mW	0-Lead QFN Lead Temperature (105)+300 C

Note 1: For UCSP solder profile information, visit www.maxim-ic.com/1st_pages/UCSP.html.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(IN = 3.8V, $\overline{\text{SHDN}}$ = IN, C_{BP} = 10nF (MAX8531), T_{A} = -40°C to +85°C, unless otherwise noted. Typical values are at T_{A} = +25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
Input Voltage	VIN					6.5	V	
Undervoltage Lockout Threshold	Vuvlo	IN rising, hysteresis is 40mV (typ)			2.25	2.42	V	
		$T_A = +25$ °C, $I_{OUT1} = I_{OUT2} = 1$ m	ıA	-1		+1		
Output Voltage Accuracy		$T_A = -40$ °C to +85°C, $I_{OUT1} = I_{OUT2} = 1$ mA $T_A = -40$ °C to +85°C, $I_{OUT1} = 0.1$ mA to 200mA, $I_{OUT2} = 0.1$ mA to 150mA		-2		+2	%	
Catput Voltage / toodracy				-3		+3		
Marriago um Orotanot Commont	1	OUT1		200			Λ	
Maximum Output Current	IOUT_	OUT2		150			mA	
Current Limit	1	OUT1		210	330	550		
Current Limit	I _{LIM}	OUT2		165	280	500	mA	
Crown of Courses	l-	No load			130	220		
Ground Current	IQ	I _{OUT1} = I _{OUT2} = 100mA		150		μA		
Dropout Voltage	V _{OUT} V _{IN}	I _{OUT} = 100mA (Note 3)			100	200	mV	
Line Regulation	ΔV_{LNR}	$IN = (OUT_ + 0.1V) \text{ to } 3.8V$	-0.2		+0.2	%/V		
Outrant Valtage Naige		10Hz to 100kHz, C _{OUT} = 10µF,	MAX8530		320		\/=	
Output Voltage Noise		I _{OUT} = 10mA	MAX8531		40		μV _{RMS}	
Discola Dais etias	DODD	10011- 1 20 1	MAX8530		60		4D	
Ripple Rejection	PSRR	100Hz, I _{OUT} = 30mA	MAX8531		62		dB	
SHUTDOWN								
OUDM Consists Comment		SHDN = 0, T _A = +25°C		0.01	1	^		
SHDN Supply Current	I _{OFF}	SHDN = 0, T _A = +85°C		0.1		μΑ		
CLIDAL languat Threadh ald	VIH	Input high voltage		1.6				
SHDN Input Threshold	VIL	Input low voltage				0.4	V	
CUDN Input Rica Current			T _A = +25°C		0.7	100	- Λ	
SHDN Input Bias Current	ISHDN	SHDN = IN or GND	T _A = +85°C		0.8		nA	
THERMAL PROTECTION								
Thermal-Shutdown Temperature	T _{SHDN}	T _J rising			160		°C	
Thermal-Shutdown Hysteresis	ΔT_{SHDN}				10		°C	

ELECTRICAL CHARACTERISTICS (continued)

(IN = 3.8V, \overline{SHDN} = IN, C_{BP} = 10nF (MAX8531), T_A = **-40°C to +85°C**, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

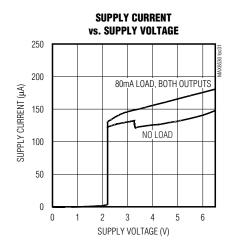
PARAMETER	SYMBOL	CONDITIONS			TYP	MAX	UNITS
RESET OUTPUT (MAX8530 ONLY)							
RESET Threshold	V _{TH}	Percentage of nominal output, O when RESET falls.	84	87	89	%	
RESET Threshold Hysteresis					0.5		%
RESET Output Voltage Low	Vol	IRESET = 500μA, IN = 3.8V			10	20	mV
(Open-Drain, Active Low)	VOL	$\overline{\text{IRESET}} = 100\mu\text{A}, \text{IN} = 1.2\text{V}$	IRESET = 100µA, IN = 1.2V		10	40	1111
RESET Open-Drain		$V_{\overline{RESET}} = 6.5V$ $T_{A} = +25^{\circ}C$ $T_{A} = +85^{\circ}C$			0.001	1	T
Output Leakage Current					0.01		μA
RESET Active Timeout Delay		From OUT1 rising to RESET rising.		100	200	360	ms

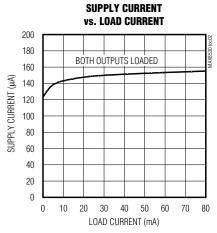
Note 2: All units are 100% production tested at $T_A = +25$ °C. Limits over the operating temperature range are guaranteed by design.

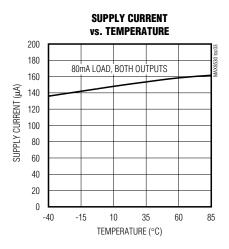
Note 3: The dropout voltage is defined as V_{IN} - V_{OUT}, when V_{OUT} is 100mV below the value of V_{OUT} for V_{IN} = V_{OUT} + 0.5V. Specification applies only when V_{OUT} ≥ 2.5V.

Typical Operating Characteristics

 $V_{OUT1} = 3.3V$, $V_{OUT2} = 1.5V$, load = 80mA (both output voltages), $V_{IN} = 3.8V$, $C_{OUT} = 2.2\mu F$, $C_{BP} = 0.01\mu F$, $C_{IN} = 2.2\mu F$, and $T_{A} = +25^{\circ}C$, unless otherwise noted.

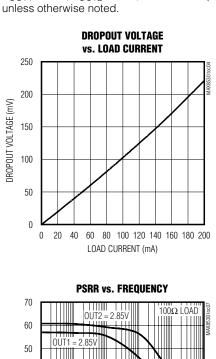


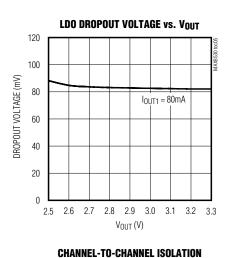


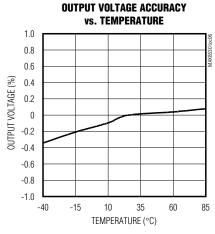


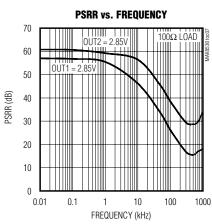
Typical Operating Characteristics (continued)

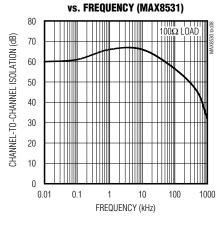
 $V_{OUT1} = 3.3V$, $V_{OUT2} = 1.5V$, load = 80mA (both output voltages), $V_{IN} = 3.8V$, $C_{OUT} = 2.2\mu$ F, $C_{BP} = 0.01\mu$ F, $C_{IN} = 2.2\mu$ F, and $T_{A} = +25^{\circ}$ C, unless otherwise noted

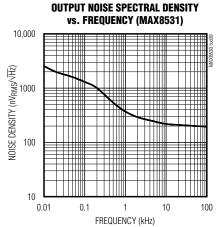


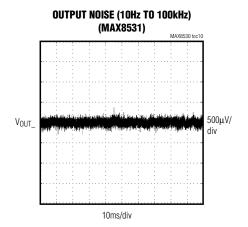


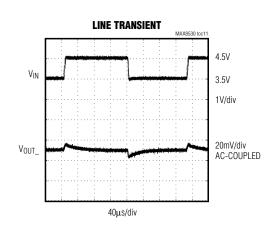






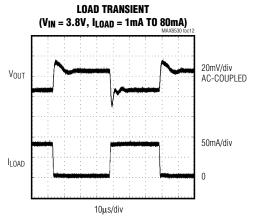


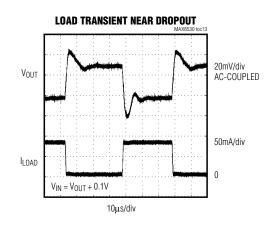


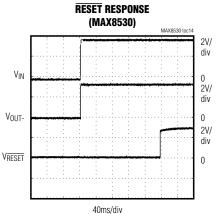


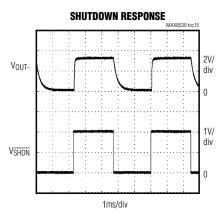
Typical Operating Characteristics (continued)

 $V_{OUT1} = 3.3V$, $V_{OUT2} = 1.5V$, load = 80mA (both output voltages), $V_{IN} = 3.8V$, $C_{OUT} = 2.2\mu F$, $C_{BP} = 0.01\mu F$, $C_{IN} = 2.2\mu F$, and $T_{A} = +25^{\circ}C$, unless otherwise noted.





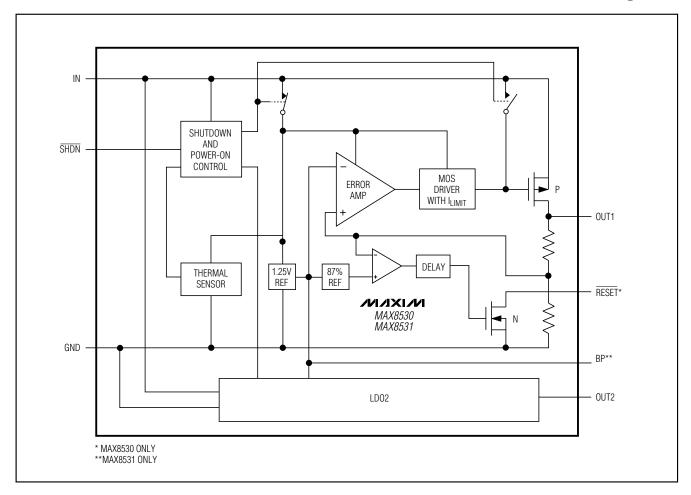




Pin Description

	PI	N				
MAX8530 (QFN)	MAX8531 (QFN)	MAX8530 (UCSP)	MAX8531 (UCSP)	NAME	FUNCTION	
1	1	В3	В3	OUT2	Regulator 2 Output. Guaranteed 150mA output current.	
2	2	B2	B2	IN	Regulator Input	
3	3	B1	B1	OUT1	Regulator 1 Output. Guaranteed 200mA output current.	
4	4	A1	A1	GND	Ground. Also functions as a heatsink for the 6-pin QFN. Solder to a large pad or the circuit board ground plane to maximize thermal dissipation.	
5	5	A2	A2	SHDN	Shutdown Input. A logic low shuts down both regulators. Connect to IN for normal operation.	
6	_	АЗ	_	RESET	Reset Active-Low Open-Drain Output. Minimum timeout of 100ms. RESET is low in shutdown.	
_	6	_	А3	BP	Reference Noise Bypass. Connect to a 0.01µF ceramic capacitor for reduced noise at both outputs.	

Functional Diagram



Detailed Description

The MAX8530/MAX8531 are low-power, low-dropout, low-quiescent current linear regulators with reset designed primarily for battery-powered applications. For preset output voltages, refer to the *Output Voltage Selector Guide*. Other combinations between 1.5V and 3.3V are available in 50mV increments. This device supplies loads up to 200mA for OUT1 and 150mA for OUT2. The MAX8530/MAX8531 consist of a 1.25V reference, error amplifiers, P-channel pass transistors, reset block, and internal feedback voltage-dividers.

The 1.25V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the

reference voltage, the pass-transistor gate is pulled lower, allowing more current to pass to the output and increasing the output voltage. If the feedback voltage is too high, the pass-transistor gate is pulled up, allowing less current to pass to the output. The output voltage is fed back through an internal resistor voltage-divider connected to the OUT_pin.

RESET

The reset circuit is active at power-up and power-down. RESET is held low at power-up. The reset timing starts once the OUT1 voltage reaches 87% (typ) of its regulation voltage. The RESET signal goes high 200ms (typ) after the OUT1 voltage reaches 87% (typ) of its regulation voltage. At power-off, RESET goes low when the OUT1 voltage is below 87% (typ) of its regulation voltage.

Shutdown

The MAX8530/MAX8531 have a single shutdown control input (SHDN). Drive SHDN low to shut down both outputs, reducing supply current to 10nA. Connect SHDN to a logic high, or IN, for normal operation.

Internal P-Channel Pass Transistor

The MAX8530/MAX8531 feature two 1Ω P-channel MOSFET pass transistors. A P-channel MOSFET provides several advantages over similar designs using PNP pass transistors, including longer battery life. It requires no base drive, substantially reducing quiescent current. PNP-based regulators waste considerable current in dropout when the pass transistor saturates and also use high base-drive currents under heavy loads. The MAX8530/MAX8531 do not suffer these problems and consume only 150µA of quiescent current whether in dropout, light-load, or heavy-load applications (see the Typical Operating Characteristics). Whereas a PNP-based regulator has dropout voltage that is independent of the load, a P-channel MOSFET's dropout voltage is proportional to load current, providing for low dropout voltage at heavy loads and extremely low dropout voltage at lighter loads.

Current Limit

The MAX8530/MAX8531 contain two independent current limiters, one for each regulator, which monitor and control the pass transistor's gate voltage and limit the output currents to 210mA and 165mA minimum. The output can be shorted to ground for an indefinite time without damaging the part.

Thermal-Overload Protection

Thermal-overload protection limits total power dissipation in the MAX8530/MAX8531. When the junction temperature exceeds $T_J = +160^{\circ}\text{C}$, the thermal sensor signals the shutdown logic, turning off the pass transistor and allowing the IC to cool. The thermal sensor turns the pass transistor on again after the IC's junction temperature cools by 10°C , resulting in a pulsed output during continuous thermal-overload conditions.

Thermal-overload protection is designed to protect the MAX8530/MAX8531 in the event of fault conditions. For continual operation, do not exceed the absolute maximum junction temperature rating of $T_J = +150^{\circ}C$.

Operating Region and Power Dissipation

The MAX8530/MAX8531s' maximum power dissipation depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the rate of airflow. The power dissipation across the device is $P = I_{OUT}$ ($V_{IN} - V_{OUT}$). Maximum power dissipation:

$$P_{MAX} = (T_J - T_A)/(\theta_{JB} + \theta_{BA})$$

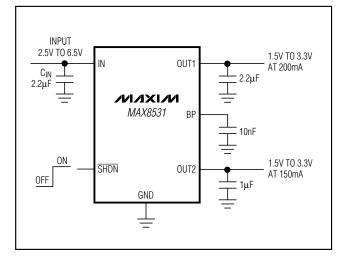
where T_J - T_A is the temperature difference between the MAX8530/MAX8531 die junction and the surrounding air, θ_{JB} (or θ_{JC}) is the thermal resistance of the package, and θ_{BA} is the thermal resistance through the printed circuit board, copper traces, and other materials to the surrounding air.

The GND pin of the MAX853_ETT__ (6-lead QFN) performs the dual functions of providing an electrical connection to the ground and channeling heat away. Connect the GND pin and exposed pad to ground using a large pad or ground plane.

Low-Noise Operation (MAX8531)

An external $0.01\mu F$ bypass capacitor at BP, in conjunction with an internal resistor, creates a lowpass filter (see the *Typical Application Circuit*). The MAX8531 exhibits $40\mu V_{RMS}$ output voltage noise with $C_{BP} = 0.01\mu F$ and $C_{OUT} = 2.2\mu F$ (see the Output Noise Spectral Density vs. Frequency graph in the *Typical Operating Characteristics*).

MAX8531 Typical Appication Circuit



Applications Information

Capacitor Selection and Regulator Stability

Use a 2.2 μ F capacitor on the MAX8530/MAX8531s' inputs. Larger input capacitor values with lower ESRs provide better supply-noise rejection and line-transient response. To reduce noise and improve load transients, use large-output capacitors, up to 10 μ F. For stable operation over the full temperature range and with rated maximum load currents, use a minimum of 2.2 μ F (or 1 μ F for <150mA loading for OUT1) and 1 μ F for OUT2.

Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. With dielectrics such as Z5U and Y5V, it is necessary to use 4.7µF or more to ensure stability at temperatures below -10°C. With X7R or X5R dielectrics, 2.2µF is sufficient at all operating temperatures. These regulators are optimized for ceramic capacitors. Tantalum capacitors are not recommended.

PSRR and Operation from Sources Other than Batteries

The MAX8530/MAX8531 is designed to deliver low dropout voltages and low quiescent currents in battery-powered systems. Power-supply rejection is 60dB at low frequencies (see the Power-Supply Rejection Ratio vs. Frequency graph in the *Typical Operating Characteristics*).

When operating from sources other than batteries, improve supply-noise rejection and transient response by increasing the values of the input and output bypass capacitors and through passive filtering techniques.

Load-Transient Considerations

The MAX8530/MAX8531 load-transient response graphs (see the *Typical Operating Characteristics*) show two components of the output response: a DC shift in the output voltage because of the different load

currents, and the transient response. Increase the output capacitor's value and decrease its ESR to attenuate transient spikes.

Input/Output (Dropout Voltage)

A regulator's minimum input/output voltage differential (or dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage. Because the MAX8530/MAX8531 use a P-channel MOSFET pass transistor, their dropout voltage is a function of drain-to-source on-resistance (RDS(ON)) multiplied by the load current (see the *Typical Operating Characteristics*).

Calculating the Maximum Output Power in UCSP

The maximum output power of the MAX8530/MAX8531 can be limited by the maximum power dissipation of the package. Obtain the maximum power dissipation by calculating the power dissipation of the package as a function of the input voltage, output voltage, and output currents. The maximum power dissipation should not exceed the package's maximum power rating:

 $P = (V_{IN(MAX)} - V_{OUT1}) \times I_{OUT1} + (V_{IN(MAX)} - V_{OUT2}) \times I_{OUT2}$

where:

V_{IN(MAX)} = Maximum input voltage

P_{MAX} = Maximum power dissipation of the package (308mW for UCSP and 1951mW for the QFN package)

V_{OUT1} = Output voltage of OUT1

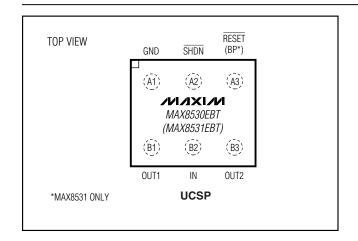
V_{OUT2} = Output voltage of OUT2

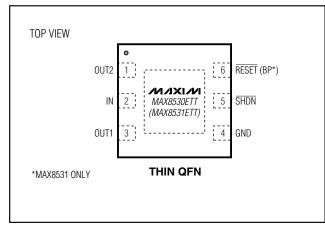
IOUT1 = Maximum output current of OUT1

IOUT2 = Maximum output current of OUT2

P should be less than $P_{\mbox{\scriptsize MAX}}.$ If P is greater than $P_{\mbox{\scriptsize MAX}},$ consider using the QFN package.

Pin Configurations





Output Voltage Selector Guide

PART	V _{OUT1}	V _{OUT2}	TOP MARK
MAX8530EBTJ2	2.85	1.8	ACR
MAX8530EBTJO	2.85	2.6	ACA
MAX8530EBTKG	2.8	3.0	ACT
MAX8530ETTP2	2.5	1.8	AET
MAX8530ETTO2	2.6	1.8	AES
MAX8530ETTK2	2.8	1.8	AER
MAX8530ETTKO	2.8	2.6	AEQ
MAX8531EBTJJ	2.85	2.85	ACG
MAX8531EBTGG	3.0	3.0	ACI
MAX8531ETTGG	3.0	3.0	AEF

Note: Contact the factory for other output voltages between 1.5V and 3.3V. The minimum order quantity is 25,000 units.

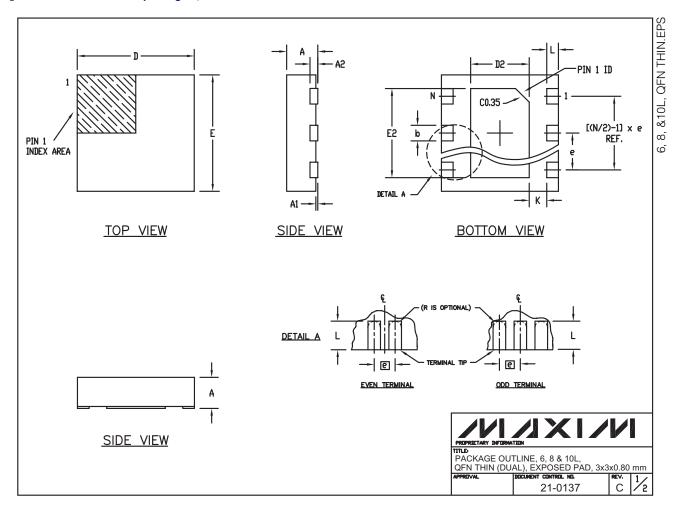
Chip Information

TRANSISTOR COUNT: 1720

PROCESS: BICMOS

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)



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Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)

COMMON DIMENSIONS					
SYMBOL	MIN.	MAX.			
Α	0.70	0.80			
D	2.90	3.10			
E	2.90	3.10			
A1	0.00	0.05			
L	0.20	0.40			
k	0.25 MIN				
A2	0.20 REF.				

PACKAGE VARIATIONS								
PKG. CODE	N	D2	E2	е	JEDEC SPEC	b	[(N/2)-1] x e	
T633-1	6	1.50-0.10	2.30-0.10	0.95 BSC	MO229 / WEEA	0.40-0.05	1.90 REF	
T833-1	8	1.50-0.10	2.30-0.10	0.65 BSC	MO229 / WEEC	0.30-0.05	1.95 REF	
T1033-1	10	1.50-0.10	2.30-0.10	0.50 BSC	MO229 / WEED-3	0.25-0.05	2.00 REF	

- NOTES:
 1. ALL DIMENSIONS ARE IN mm. ANGLES IN DEGREES.
 2. COPLANARITY SHALL NOT EXCEED 0.08 mm.
 3. WARPAGE SHALL NOT EXCEED 0.10 mm.

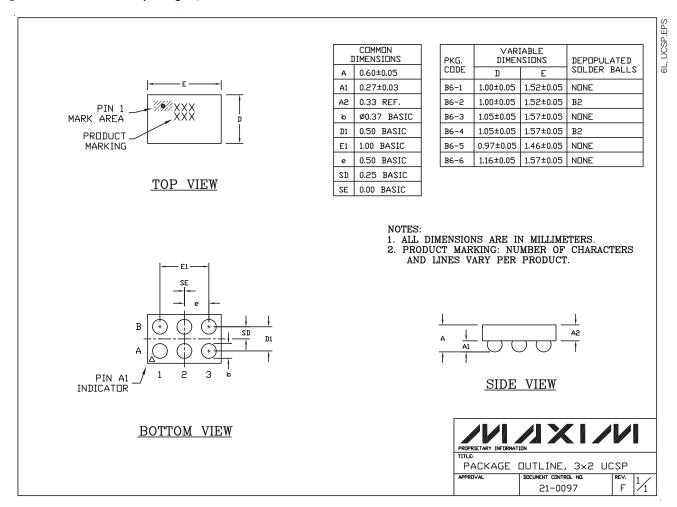
- 4. PACKAGE LENGTH/PACKAGE WIDTH ARE CONSIDERED AS SPECIAL CHARACTERISTIC(S).

 5. DRAWING CONFORMS TO JEDEC MO220.



Package Information (continued)

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